



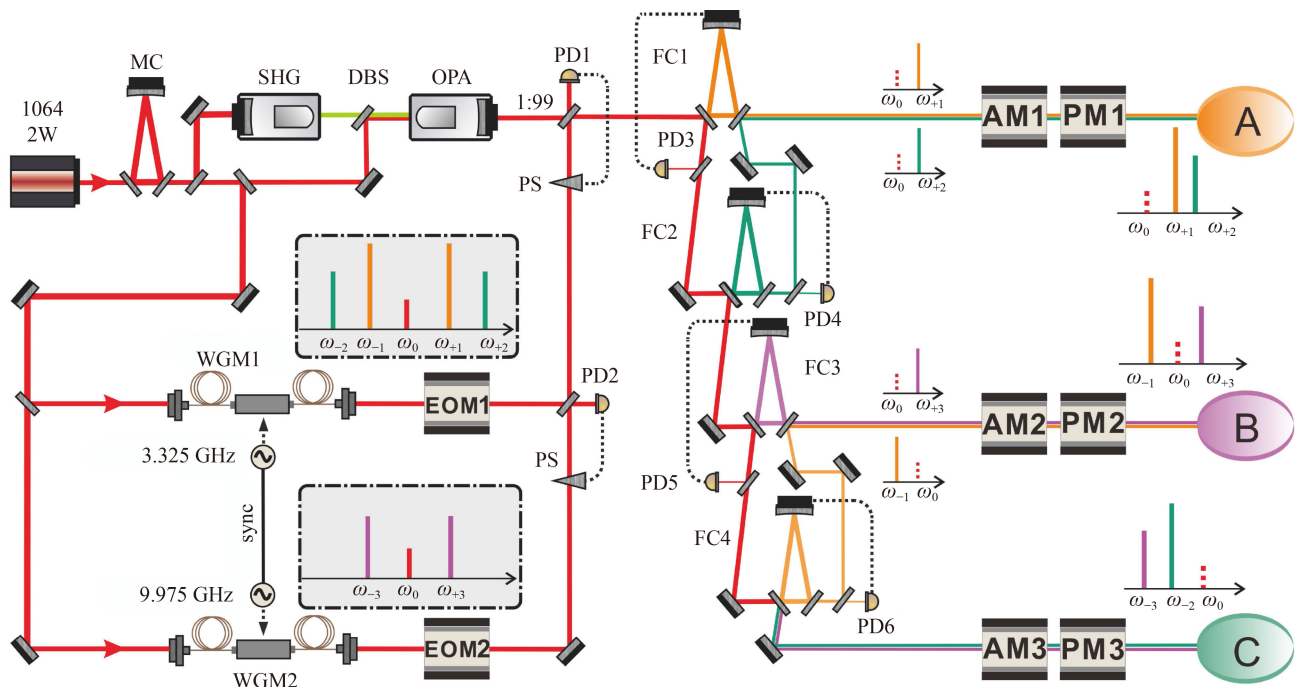
that establish a temporary quantum link at a time. But the secure channel capacity of each link is limited due to time-sharing configuration. The third is multi-partite (high-dimensional) entanglement based architecture [18–22], which is established by sharing entanglement state among several parties. However, in the configuration the node number is dependent how many partite is the entanglement state, making the architecture lack of flexible and scalable. The last is fully-connected networks, in which every party is directly connected to every other party by sharing bipartite entanglement between any two parties. Following the thought, fully-connected network architectures were demonstrated, exploiting a polarization-entangled photon source as quantum resource. Continuous variable (CV) physical system enjoys deterministic quantum state generation, unconditional operations and ultrahigh detection efficiencies [23–26], which represents the complementary relation with respect to discrete-variable system. However, fully-connected network architecture with more than two parties for a CV optical system remains experimentally unexplored due to challenges in implementing demultiplexing and multiplexing operation with ultra-low losses.

In the previous presentation, we demonstrated four-fold channel multiplexing point-to-point quantum communication based on CV [27], exploiting a squeezed state of light via low-loss demultiplexing operation and high-performance feedback control. Following the previous presentation, we report the demonstration of a fully-connected CV quantum dense coding communication

network with high capacity of each link and low crosstalk between two links. Three pairs of entangled sideband modes from a squeezed field are spatially separated by demultiplexing operation, then recombining into new group according to network requirement, distributing to each party connected with the network. In this way, we distribute bipartite entanglement between any pairs of parties. In virtue of a frequency-comb-type control scheme, all of these ring filter cavities (as a frequency-dependent beam splitter) and relative phases operate stably at the theoretical working point, constructing general and reliable demultiplexing and multiplexing system. The demonstration opens an innovative possibility to implement more large-scale quantum communication network.

## 2 Experimental setup

The schematic diagram of the fully-connected quantum dense coding communication network is shown in Fig. 1. The main laser source is home-made single-frequency laser with the wavelength and output power of 1064 nm and 2 W. The laser transmits through a mode cleaner (MC) that provides spatial-temporal filtering, polarization purifying, intensity-noise and phase-noise suppressing for the downstream experiment, generating the output power of 1.6 W. Approximately 1.4 W of the output power is injected into the second harmonic generator (SHG) where the pump beam is generated for the optical



**Fig. 1** Schematic of the experimental setup. MC, mode cleaner; SHG, second harmonic generator; OPA, optical parametric amplifier; FC, filter cavity; WGM, waveguide electro-optic modulator; EOM, electro-optic phase modulator; PS, phase shifter; DBS, dichroic beam splitter; PD, photodetector; AM, amplitude modulator; PM, phase modulator.

parametric amplifier (OPA) downstream. A small fraction of the output beam from the MC is split off and serves as the seed beam for OPA, auxiliary control beam generation and local oscillator generation. Utilizing an electro-optic modulator (EOM), a phase modulation signal is imprinted on the laser beam, the length of the MC and SHG is stabilized to meet the resonance condition for the laser beam via the Pound–Drever–Hall locking technique [28]. For simplicity, the control loops are omitted in Fig. 1.

Our OPA is a semi-monolithic cavity consisting of a piezo-actuated concave mirror and a periodically poled  $\text{KTiOPO}_4$  (PPKTP) crystal with the dimensions of  $1 \text{ mm} \times 2 \text{ mm} \times 10 \text{ mm}$ . In order to lengthen the effective nonlinear interaction, the crystal end face has a radius of curvature of 12 mm with high reflectivity (HR) at 1064 nm and high transmission at 532 nm, thus serving as one of cavity mirror. The plane front face of the crystal has an antireflective (AR) coating for both wavelengths. We set an air gap of 27 mm length between the crystal and the concave mirror to optimize the waist size inside the crystal. The concave mirror with a radius of curvature of 30 mm has a transmissivity of  $12\% \pm 1.5\%$  for 1064 nm and HR for 532 nm, which is used as the output coupler. Such a design constructs a single-resonator optical cavity with high escape efficiency [29, 30]. The OPA operates at parametric deamplification state, suppressing the noise of amplitude component to below the shot noise limit (SNL). The squeezed field involves a lot of Einstein–Podolsky–Rosen (EPR) entangled modes that distribute symmetrically around the half pump frequency [31]. Each pair of EPR entangled modes have a bandwidth of approximately 70 MHz, and frequency separation between two adjacent modes of 3.325 GHz that corresponds to free spectral range of the OPA [32–34]. Therefore, in virtue of the quantum resource, a quantum communication network with high channel capacity and low crosstalk has the potential to be constructed.

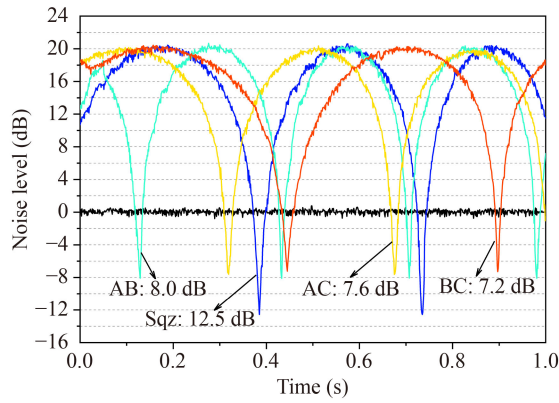
A cascade of four ring filter cavities (FC) is employed as the bandpass filters to spatially separate and recombine these sideband modes. The FC consists of two plane mirrors with a transmissivity of 10% for s-polarized beam and one curved mirror with high reflectivity. In order to efficiently separate these sideband modes, the four FCs in turn have a cavity length of 165 mm, 165 mm, 190 mm and 165 mm, respectively. Four FCs in turn resonant with these sideband modes at the frequencies  $\omega_{+1}$ ,  $\omega_{+2}$ ,  $\omega_{+3}$ ,  $\omega_{-1}$ , respectively, thus a corresponding sideband mode is transmitted from a FC, and the other sideband modes are reflected. The resonance condition of each FC is guaranteed by utilizing an active control loop, in which an auxiliary control beam is employed to extract the error signal. The auxiliary control beam that includes these coherent sidebands  $\omega_{\pm 1}$ ,  $\omega_{\pm 2}$ ,  $\omega_{\pm 3}$ , is generated by using two synchronous waveguide electro-optic

modulators (WGM1 and WGM2) with the driving frequency of 3.325 GHz and 9.975 GHz. The sideband modes  $\omega_{+2}$  separated from the FC2 is coupled with  $\omega_{+1}$  on the output mirror of the FC1, recombining the sideband modes  $\omega_{+2}$ ,  $\omega_{+1}$  into a single physical path, in which the modes  $\omega_{+2}$ ,  $\omega_{+1}$  are distributed to the party A. Similar to the previous operation, the sideband mode  $\omega_{-3}$  is multiplexed with  $\omega_{-1}$ , and then distributed to the party B. The remaining sideband modes  $\omega_{-2}$ ,  $\omega_{-3}$  reflected from FC4 are routed to the party C. Each party now has two quantum channels, thus sharing a different entangled state with every other party. According to the standard technical protocol of CV quantum dense coding communication [35, 36], we encode the classical amplitude and phase signals on one half of each EPR beam respectively by electro-optic amplitude modulators (AM) and electro-optic phase modulators (PM), which represents the transmitted signals in quantum communication. The signals are decoded at the receiver with the aid of the other half of each EPR beam.

For implementing decoding operation, three pairs of local oscillators (LOs) at  $\omega_{\pm 1}$ ,  $\omega_{\pm 2}$ ,  $\omega_{\pm 3}$  have to be prepared as reference beam of balanced homodyne detections (BHDs). These LOs are generated by three WGMs and subsequent optical mode cleaners, which has been detailedly shown in our previous publications [27]. Each sideband mode is carefully mode matched with the corresponding frequency LO on a 50/50 beam splitter, and the output of the 50/50 beam splitter is directed toward a BHD to observe noise level. The BHDs, with the common mode rejection ratio of 75 dB [37], is built from a pair of p-i-n photodiodes (from Laser Components) with quantum efficiency of more than 99%.

### 3 Experimental results

Figure 2 presents the measured results of original squeezed degree with Blue line after the OPA but before the demultiplexing. The directly observed squeezing level is 12.5 dB below the SNL. To characterize the performance of the sideband entangled states after the demultiplexing and recombining operation, the measurement is performed directly without the AM and PM presenting. With the signal ports of the two BHDs blocked, the output of the joint measurement corresponds to the SNL (Black line). One of the BHD's phase keeps constant while another BHD's phase is scanned linearly in time, the spectral densities of the joint measurement are recorded. The variances of amplitude sum and phase difference are unbiased and equal to 8.0 dB for the communication link AB (Green line,  $\omega_{\pm 1}$ ), 7.6 dB for the communication link AC (Yellow line,  $\omega_{\pm 2}$ ) and 7.2 dB for the communication link BC (Red line,  $\omega_{\pm 3}$ ), respectively, without the correct of the electronic noise. According to the inseparable criteria [38] of

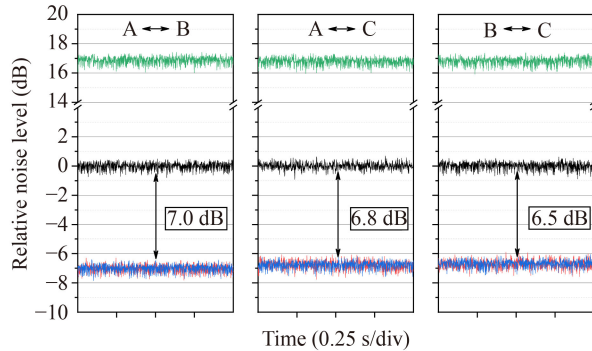


**Fig. 2** Balanced homodyne measurements of the quadrature noise variances. The measurement is recorded at the analysis frequency of 2 MHz; RBW, resolution bandwidth = 300 kHz; VBW, video bandwidth = 200 Hz.

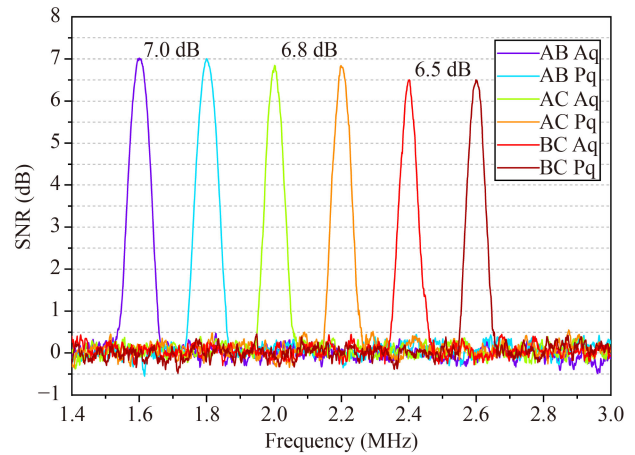
$\text{Var}(\hat{X}_A + \hat{X}_B) + \text{Var}(\hat{Y}_A - \hat{Y}_B) < 2$ , the measured correlation variances in the left of the inequality are 0.317 for  $\omega_{\pm 1}$ , 0.348 for  $\omega_{\pm 2}$ , 0.381 for  $\omega_{\pm 3}$ , respectively, witnessing the presence of entanglement of these sideband modes.

Further, several AMs and PMs are placed in each communication link to simulate the transmitted signals. Without imposing modulation signal on the AMs and PMs, we record the noise power spectrum at three communication links, as shown in Fig. 3. The green line represents channel background noise of one half of the EPR beam without the help of the other half of the EPR beam. The red and blue lines are the noise spectrum of amplitude sum and phase difference, which is 7.0 dB for the communication link AB, 6.8 dB for the communication link AC, 6.5 dB for the communication link BC less than the SNL.

We simulate the transmitted signal with the intensity that exactly equals to the SNL, the signal-to noise ratios (SNRs) at different links are shown in Fig. 4. It can be



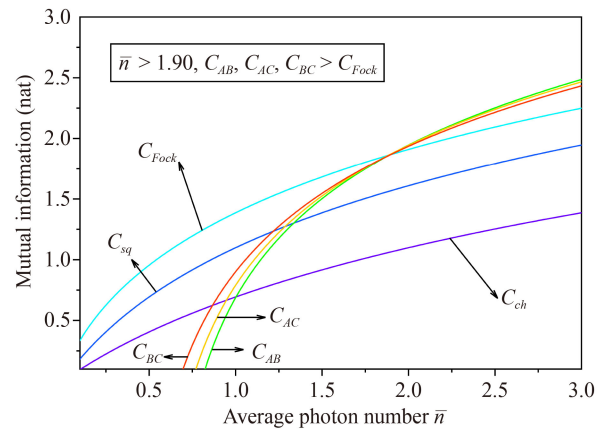
**Fig. 3** Normalized quantum noise level of amplitude sum and phase difference at three communication links, corresponding to three pairs of entangled sideband modes shared by each two parties. When one half of the EPR beam is blocked, the noise spectrum is higher than the SNL.



**Fig. 4** Normalized SNR of each communication link based on QDC protocol. Aq, amplitude quadrature; Pq, phase quadrature.

seen that the sensitivities on both the amplitude and phase quadratures are beyond that of the SNL by means of exploiting the EPR entanglement, while the signal is submerged in the background noise without the EPR entanglement present. In order to clearly present the measurement results, we use different modulation frequency 1.6 MHz (amplitude quadrature) and 1.8 MHz (phase quadrature) at the communication link AB, 2.0 MHz (amplitude quadrature) and 2.2 MHz (phase quadrature) at the communication link AC, 2.4 MHz (amplitude quadrature) and 2.6 MHz (phase quadrature) at the communication link BC to simulate the transmitted signals, corresponding to the SNR of 7.0 dB, 6.8 dB, 6.5 dB, respectively. For each link, the SNR that retrieves the signals is equal for amplitude and phase quadratures.

According to communication theory, the signal-to noise ratio exactly is an indicator of channel capacity [39–41]. Using the above measured SNRs, we evaluate



**Fig. 5** Channel capacity by various channels, measured by mutual information as functions of the average photon number.



the channel capacity of each communication link at quantum dense coding communication network. Figure 5 compares the channel capacity of each link with that of other communication protocols. We can see that for communication link BC (AB), the channel capacity of the quantum dense coding communication  $C_{BC}$  ( $C_{AB}$ ) has exceeded that of other CV quantum communication protocols, such as coherent state  $C_{ch}$ , squeezed state with the optimal value  $C_{sq}$ , Fock state communication  $C_{Fock}$ , which is the Holevo limit of a single-mode bosonic channel, under the condition of  $> 0.87$  (1.01), 1.22 (1.34), 1.90 (1.90) ( $\bar{n}$  stands for the average photon number), respectively [41].

In the above experiment, in order to facilitate the exhibition of the communication process between each user, the amplitude and phase quadrature are demonstrated using quantum dense coding with a small frequency difference (0.2 MHz), meanwhile employing different frequency point in each communication link. Limited by the bandwidth of OPA and FC, the bandwidth of each communication link are dozens of MHz. By expanding the bandwidth of each channel and changing the dense coding frequency, combined with other quantum communication protocols, we hope to achieve continuous variable quantum communication networks with larger communication capacity.

## 4 Conclusion

In conclusion, we have successfully realized a proof-of-principle demonstration of a CV quantum dense coding communication network. Any party can be connected to any other party connected with the network by their own private link, which is constructed by sharing a pair of exclusive sideband entanglement modes. Therefore, all of three parties is fully connected via sideband entanglement optical modes from one squeezed state of light. Thanks to the special auxiliary control scheme, we build the low-loss demultiplexing and multiplexing system, guaranteeing high channel capacity of each link. Since CV entanglement state is susceptible to system loss, the demultiplexing and multiplexing system with low loss is extremely important to achieve the high-performance communication network. It is worthy of noting that the scheme is easily extended to communication network with more parties, if we can combine with new technology, including the filter cavity of waveguide [42], and integrate silicon-based material [43–45]. The stable control of the entire system can be achieved through passive methods such as temperature factor, greatly reducing the difficulty and complexity, and further improve the scalability of proposed network.

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